

# Three-Phase Full-Bridge Intelligent Power Module

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## 1、 Summary

The SPB15SDG is a three-phase IGBT inverter module featuring an integrated IGBT and its gate driver chip, designed for driving brushless DC motors and permanent magnet brushless motors. The gate driver chip is optimized for IGBTs to minimize switching losses and electromagnetic interference. It incorporates multiple protection functions including undervoltage lockout, overcurrent protection, and fault reporting, with an integrated thermistor for temperature monitoring. The driver circuit for the high-side IGBT employs a self-bail power supply design requiring only a single power source. Each low-side IGBT's emitter is individually routed, enabling separate sampling of three-phase currents and support for various control algorithms. The circuit is packaged in a DIP-24H package with heat sinks.

### Characteristic

600V/15A three-phase IGBT inverter module with built-in IGBT, FWD, BSD, HVIC, LVIC

The high side drive circuit uses the bootstrap method to generate a floating power supply and has a bootstrap diode with a current limiting resistor built in

The gate driver chip has a variety of protection functions, including under voltage protection, over current protection, over temperature protection

Fully compatible with the interface of 3.3V,5V and 15V MCU

The negative end of the three-phase is drawn separately for current detection

Fault status report output

Linear temperature sensing output

Encapsulation: HDIP24 with heat sink

## 2、 Function Block Diagram and Pin Description

### 2. 1. Function Block Diagram

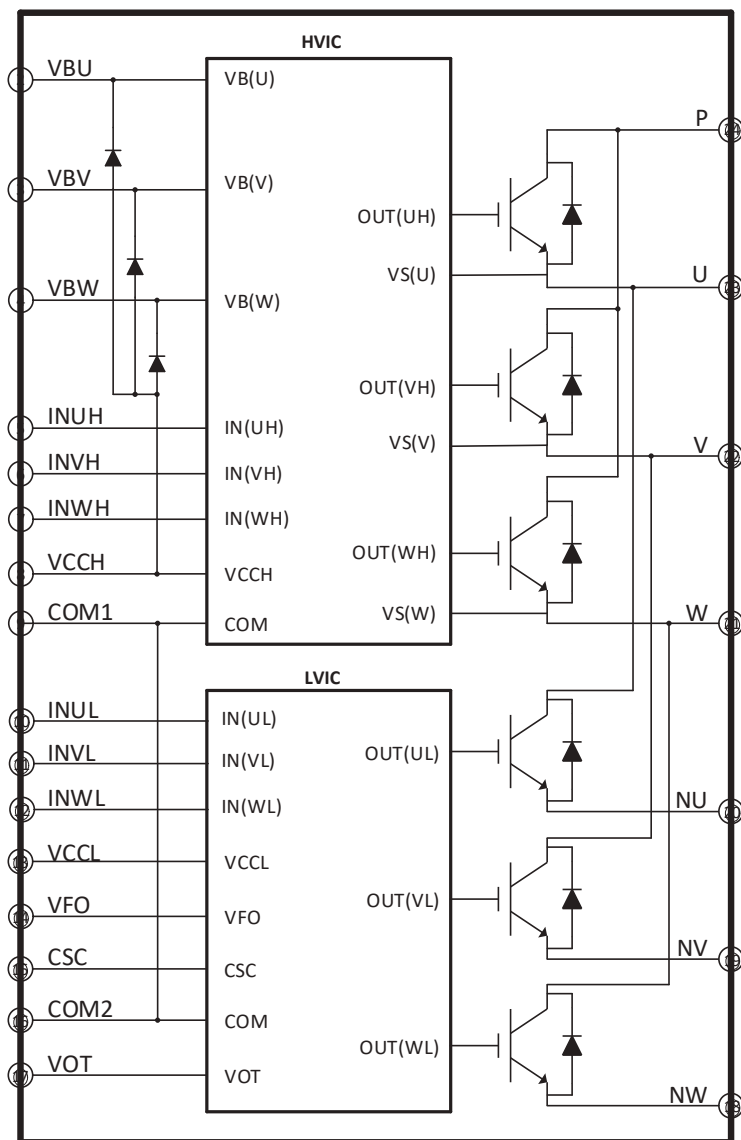


Figure 1 SPB15CDG Functional Block Diagram

### 2. 2. Functional Description

The SPB15CDG circuit integrates high-voltage ICs (HVIC), low-voltage ICs (LVIC), power devices (IGBT), freewheeling diodes (FRD), and bootstrap diodes (BSD), significantly enhancing integration density. Multiple protection features are incorporated to further improve circuit reliability. The emitter terminals of the three low-side IGBTs are separately routed, enabling independent sampling of three-phase currents for diverse drive configurations. The driver circuit for the high-side IGBT employs a bootstrap method to generate floating power supply, with the bootstrap diode integrated into the circuit requiring minimal external components. All three-phase floating power supplies incorporate undervoltage protection to prevent IGBT operation under high-power states. The SPB15CDG offers comprehensive protection mechanisms including undervoltage, overtemperature, and overcurrent safeguards. Additionally, it integrates a temperature sensing module that outputs an analog voltage proportional to temperature, facilitating MCU monitoring of IPM module thermal performance.

### 2. 3. Pin Arrangement Diagram

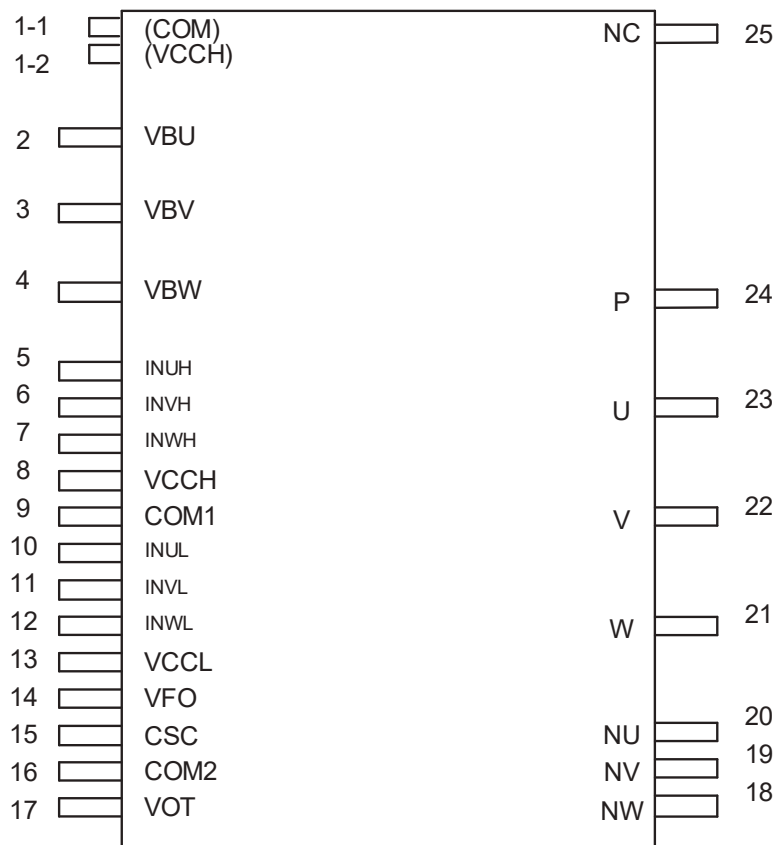


Figure 2 Pin Arrangement

### 2. 4. Pin Description

pin	Title	Description of the function of the leg
1-1	(COM)	Internal public ground terminal, no connection
1-2	(VCCH)	Internal power terminals, no connection
2	VBU	U phase high side IGBT drive suspension supply voltage
3	VBV	V phase high side IGBT drive suspension power supply voltage
4	VBW	V phase high side IGBT drive suspension supply voltage
5	INUH	U phase high side signal input
6	INVH	V Phase high side signal input
7	INWH	W phase high side signal input
8	VCCH	High side gate drive power supply voltage
9	COM1	Common to all modules
10	INUL	U phase low side signal input
11	INVL	V phase low side signal input
12	INWL	W phase low side signal input

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pin	Title	Description of the function of the leg
13	VCCL	Low side gate drive power supply voltage
14	VFO	Fault outputs
15	CSC	External capacitor, used for short circuit current detection input and low pass filter
16	COM2	Common to all modules
17	VOT	Temperature output terminal
18	NW	W phase DC negative end
19	NV	V phase DC negative end
20	NU	U phase DC negative terminal
21	W	W-phase output
22	V	V-phase output
23	U	U phase output
24	P	DC positive
25	NC	connectionless

### 3、 Electrical Characteristics

#### 3. 1. Limiting Parameters (note 1)

Parameter Name	symbol	Scope of parameters	unit
IGBT part			
supply voltage	$V_{PN}$	450	V
Power supply voltage surge voltage	$V_{PN(SURGE)}$	500	V
Cathode-emitter voltage	$V_{CES}$	600	V
The continuous collector current of a single IGBT, $T_c = 25^\circ\text{C}$ and $T_j < 150^\circ\text{C}$	$I_C$	15	A
The collector peak current of a single IGBT, $T_c = 25^\circ\text{C}$ , $T_j < 150^\circ\text{C}$ , pulse width $< 1\text{ms}$	$I_{CP}$	30	A
The maximum dissipation power rate of the collector of a single IGBT, $T_c = 25^\circ\text{C}$	$P_C$	35	W
control section			
Control the power supply voltage	$V_{CC}$	20	V
Voltage of floating power supply	$V_{BS}$	20	V
Input signal voltage	$V_{IN}$	$-0.5 \sim V_{CC} + 0.5$	V
Fault output voltage	$V_{FO}$	$-0.5 \sim V_{CC} + 0.5$	
Fault output current, VFO terminal injection current	$I_{FO}$	1	mA
Input voltage for current detection terminal	$V_{SC}$	$-0.5 \sim V_{CC} + 0.5$	V

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Parameter Name	symbol	Scope of parameters	unit
Self-举二极管部分			
Maximum repeated reverse voltage	$V_{RRM}$	600	V
Forward current, $T_c = 25^\circ\text{C}$ , $T_j < 150^\circ\text{C}$	$I_F$	0.5	A
Positive peak current, $T_c = 25^\circ\text{C}$ , $T_j < 150^\circ\text{C}$ pulse width $< 1\text{ms}$	$I_{FP}$	1.5	A
junction temperature	$T_j$	-40~150	
overall unit			
Short circuit protection limit voltage, $V_{CC} = V_{BS} = 13.5\text{V} \sim 16.5\text{V}$ , $T_j = 150^\circ\text{C}$ , single time and less than $2\mu\text{s}$	$V_{PN (PROT)}$	400	V
Working shell temperature range, $-40^\circ\text{C} \leq T_j \leq 150^\circ\text{C}$ (Note 2)	$T_c$	-20~100	
Storage temperature range	$T_{STG}$	-40~125	
IGBT shell thermal resistance	$R_{\theta JCQ}$	3.0	/W
FRD Shell thermal resistance	$R_{\theta JCF}$	3.9	/W
Insulated voltage 60Hz, sine wave, 1 minute connection of pins to heat sink	$V_{ISO}$	1500	V <sub>rms</sub>
Installation twist installation screw: -M3, recommended value 0.62N.m	T	0.5~0.8	N.m

Note 1: The maximum limit value refers to the condition beyond which the chip may be damaged. Electrical parameters define the specifications of DC and AC parameters of the device within the operating range and under test conditions that guarantee specific performance indicators;

Note 2: The maximum junction temperature of the power chip is  $150^\circ\text{C}$ . In order to ensure the safe operation of the IPM, it is recommended that the average junction temperature  $T_j \leq 130^\circ\text{C}$  ( $@T_c \leq 100^\circ\text{C}$ )

### 3. 2. Recommended Working Conditions

Parameter Name	symbol	Regulatory values			unit
		minimum	typical case	maximum	
Bus voltage between PN	$V_{PN}$	-	300	400	V
Control the power supply voltage	$V_{CC}$	13.5	15	16.5	V
High side control voltage	$V_{BS}$	13.5	15	16.5	V
Control voltage fluctuations	$dV_{CC}/dt$ $dV_{BS}/dt$	-1	-	1	V/ $\mu\text{s}$
Input turn-on threshold voltage	$V_{IN (ON)}$	3.0	-	$V_{CC}$	V
Input turn-off threshold voltage	$V_{IN (OFF)}$	0	-	0.6	V
Prevent dead time when	$T_{dead}$	1.0	-	-	$\mu\text{s}$

bridge arms are straight through					
PWM switching frequency	$f_{PWM}$	-	-	20	$\frac{KH}{z}$
COM variations (between COM $N_U$ , $N_V$ , and $N_W$ )	$V_{COM}$	-5	-	5	V

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### 3. 3. Electrical Properties

Unless otherwise specified,  $T_{amb} = 25^{\circ}\text{C}$ ,  $V_{CC} = V_{BS} = 15\text{V}$

Parameter Name	symbol	test condition	Regulatory values			unit	
			minimum	typical case	maximum		
IGBT part							
Collective collector-emitter saturation voltage	$V_{CE(SAT)}$	$V_{CC}=V_{BS}=15\text{V}$ $V_{IN}=5\text{V}$ , $I_C=15\text{A}$ $T_J=25$	-	1.8	2.3	V	
Collective collector-emitter leakage current	$I_{CES}$	$V_{CE}=V_{CES}$	-	-	1	mA	
FRD direct voltage	$V_F$	$V_{IN}=0\text{V}$ , $I_F=15\text{A}$ $T_J=25$	-	1.8	2.3	V	
switching time	High side	$t_{ON}$	-	0.95	-	μs	
		$t_C(ON)$	-	0.40	-	μs	
		$t_{OFF}$	-	0.95	-	μs	
		$t_C(OFF)$	-	0.15	-	μs	
		$t_{RR}$	$V_{PN}=300\text{V}$ $V_{CC}=V_{BS}=15\text{V}$	-	0.06	-	μs
	low side	$t_{ON}$	$I_C=15\text{A}$ $V_{IN}=0\text{V}\sim 5\text{V}$	-	0.85	-	μs
		$t_C(ON)$		-	0.40	-	μs
		$t_{OFF}$		-	0.85	-	μs
		$t_C(OFF)$		-	0.15	-	μs
		$t_{RR}$		-	0.06	-	μs
control section							
VCC static current (ON)	$I_{QCC\_ON}$	$V_{CC}=15\text{V}$ , $V_{IN}=5\text{V}$	-	-	2.8	mA	
VCC static current (OFF)	$I_{QCC\_OFF}$	$V_{CC}=15\text{V}$ , $V_{IN}=0\text{V}$	-	-	2.8	mA	
VBS quiescent current	$I_{QBS}$	$V_{CC}=V_{BS}=15\text{V}$ $V_{INH(UVW)}=0\text{V}$	-	-	80	μA	
The fault output has a high voltage	$V_{FOH}$	$V_{CSC}=0\text{V}$ , VFO pull-up 10kΩ resis-	4.9	-	-	V	

		tor to 5V				
The fault outputs a low voltage	$V_{FOL}$	$V_{CSC}=1V,$ $I_{FO}=1mA$	-	-	0.95	.
Fault outputs pulse length	$T_{FO}$	$V_{CC}=15V$	20	60	-	.
short-circuit pr- otection trigger voltage	$V_{CS}$	$V_{CC}=15V$	0.43	0.48	0.53	.

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Parameter Name	symbol	test condition	Regulatory values			unit
			minimum	typical case	maximum	
Overtemperature protection	$T_{SD}$	$V_{CC}=15V$	110	130	150	
Overtemperature protection hysteresis	$T_{SD}$	$V_{CC}=15V$	-	10	-	
Temperature outputs	$V_{OT}$	$T_A=25$	0.88	1.13	1.39	
		$T_A=90$	2.63	2.77	2.91	
VCC under-voltage protection action voltage	$UV_{CCD}$	-	10.5	11.5	12.5	
VCC under-voltage protection recovery voltage	$UV_{CCR}$	-	11.0	12.0	13.0	
VBS under-protected action voltage	$UV_{BSD}$	-	9	10	11	
VBS Under-Pressure Protection Recovery Voltage	$UV_{BSR}$	-	9.5	10.5	11.5	
Input start voltage	$V_{IH}$	$V_{CC}=15V$	-	2.4	2.9	
Input cutoff voltage	$V_{IL}$	$V_{CC}=15V$	0.8	1.3	-	
Self-oscillating diode section						
direct voltage	$V_F$	$I_F=0.1A,$ $T_C=25$	-	10.7	-	
reverse recovery time	$t_{RR}$	$I_F=0.1A,$ $T_C=25$	-	80	-	

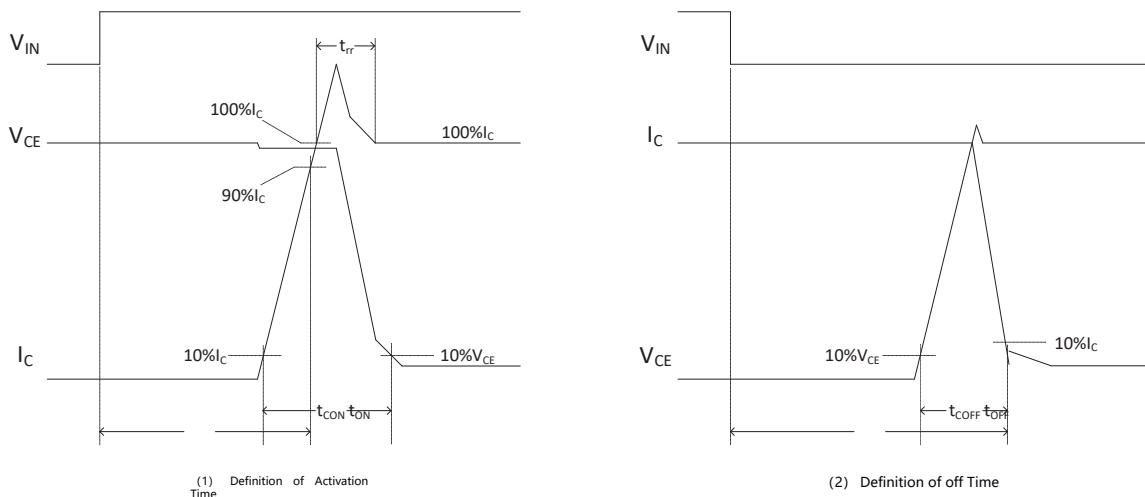


Figure 3 Switching Time Definition

IF vs. VF

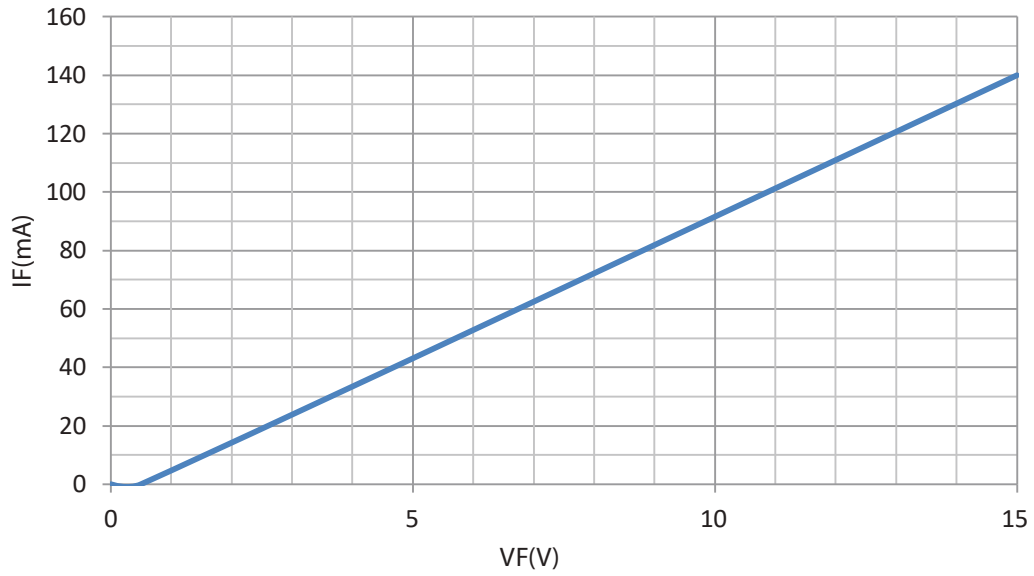


FIG. 4 Self-举二极管 Characteristic Curve

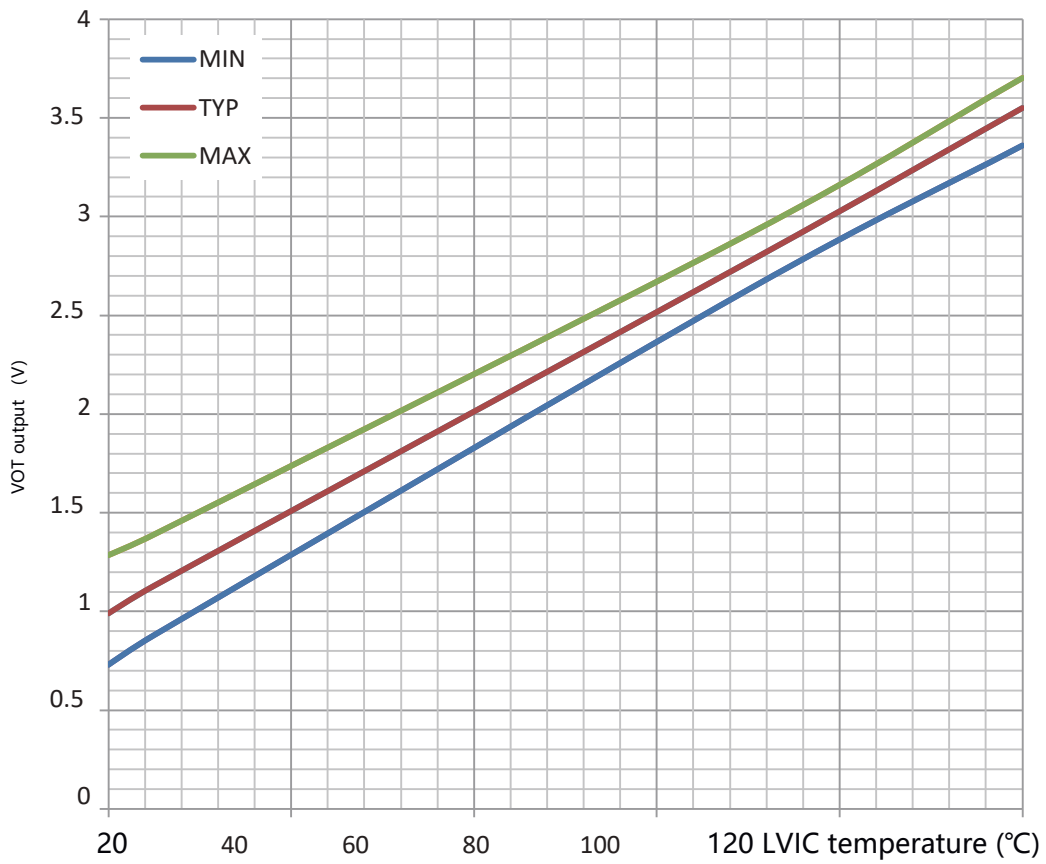


Figure 5 Output Characteristics of V<sub>OT</sub>

## 4、 Functional Description

### 4. 1. Description of Overtemperature Protection Function

The low side LVIC has overtemperature protection function, and its working sequence is as follows:

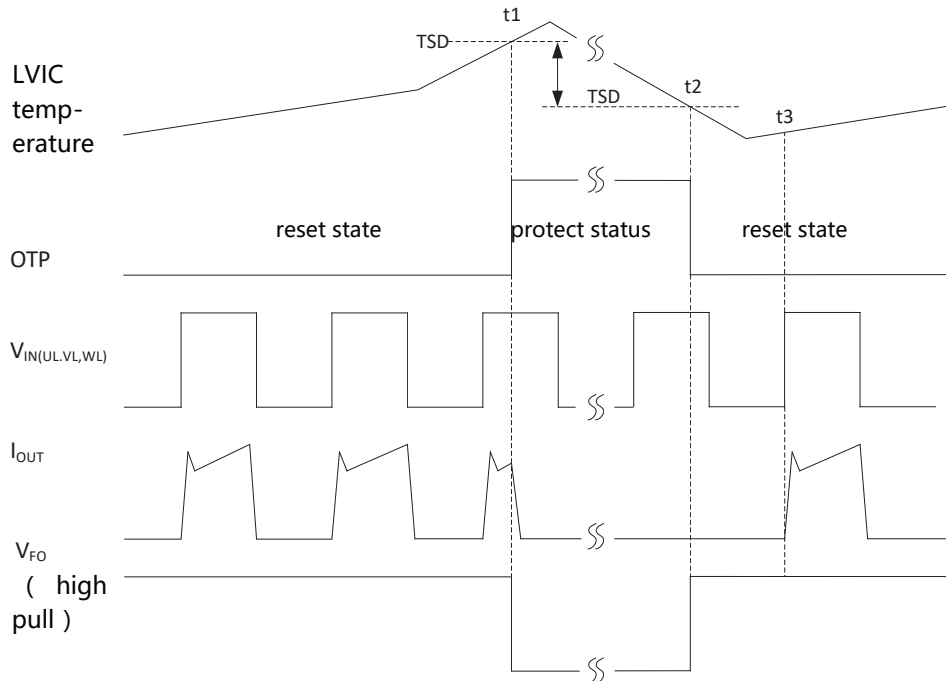


Figure 6 LVIC Overtemperature Protection Function Timing Diagram

0~t1: Before the temperature of LVIC rises to the overtemperature protection point, the circuit works normally. At this time, the current is provided to the load and  $V_{FO}$  outputs a high level, that is, the fault-free state;

t1~t2: After the temperature of LVIC rises to the overtemperature protection point, the circuit will not respond to the input signal, all low-side IGBT are turned off, and the VFO outputs a low level to report the fault status;

t2~t3: After the temperature of LVIC drops to the overtemperature protection recovery point, the circuit will not respond to the input signal immediately, but will wait for the next on signal of the input signal. At this time, the VFO outputs a high level, that is, no fault state;

t3: The circuit works normally, the IGBT is on, and the current is provided to the load.

(TSD and  $\Delta TSD$  parameter values are shown in the electrical characteristics section of Chapter 3)

## 4. 2. Description of Low Side Undervoltage Protection Function

The low side control power supply  $V_{CCL}$  has undervoltage lockout protection function, and its working timing is as follows:

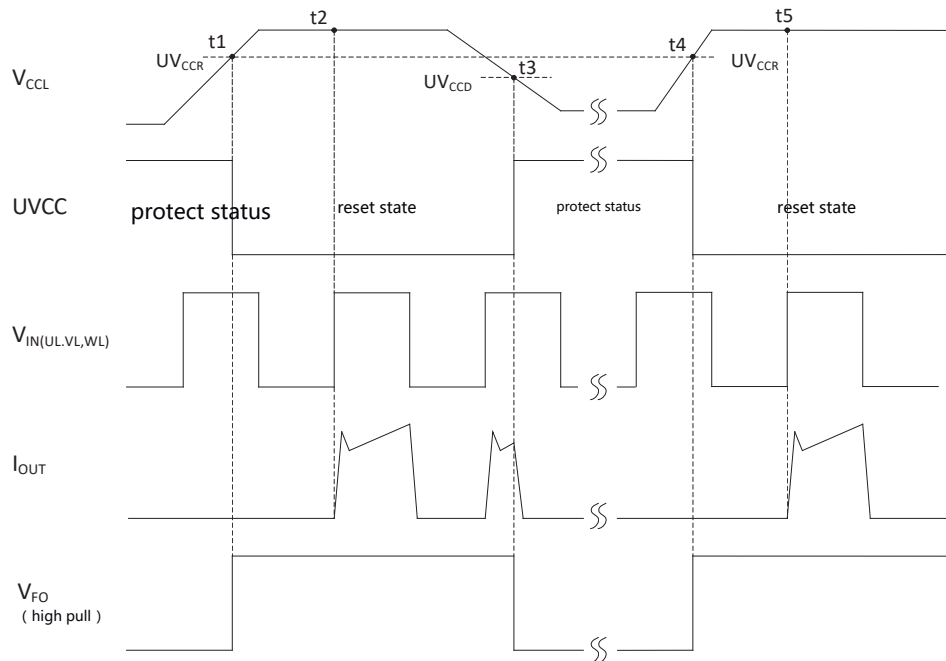


Figure 7 Schematic Diagram of Control Power Supply Under-Voltage Lockout Function

0~t1: Before  $V_{CCL}$  rises to the  $UV_{CCR}$  threshold, the UVCC is in the protection state and the circuit does not respond to the input signal. At this time,  $V_{FO}$  outputs a low level and reports the fault status;

t1~t2: When  $V_{CCL}$  rises above  $UV_{CCR}$ , the circuit does not respond to the input signal immediately, but waits for the next  $\bar{\quad}$  on signal of the input signal. At this time,  $V_{FO}$  outputs a high level, that is, no fault state;

t2~t3: the circuit works normally;

t3~t4: When  $V_{CCL}$  drops to the  $UV_{CCD}$  threshold, the circuit enters the under-voltage lockout protection state and the output is immediately turned off. In the protection state,  $V_{FO}$  outputs a low level to report the fault status;

t4: When  $V_{CCL}$  recovers to above  $UV_{CCR}$ , UVCC enters the reset state, and the circuit works normally from the next on signal of the input signal;

t5: The circuit works normally, the IGBT is on, and the current is provided to the load.

(See chapter 3, electrical characteristics section for parameter values of  $\langle UV \rangle 15$  and  $\langle UV \rangle 16$ )

### 4. 3. Description of High Side Undervoltage Protection Function

The floating power supply of the three channels on the high side has undervoltage lockout protection function (no  $V_{FO}$  fault report), and its working sequence is as follows:

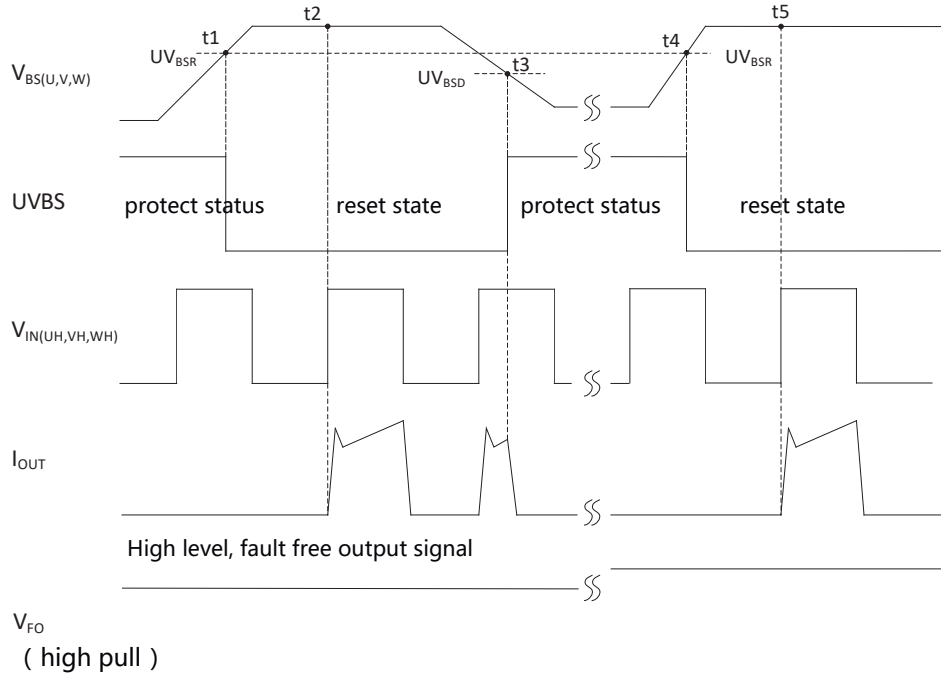


Figure 8 Floating Power Supply Under Voltage Lock Function Timing Diagram

0~t1: Before  $V_{BS}$  rises to the  $UV_{BSR}$  threshold, the UVBS is in the protection state and the circuit does not respond to the input signal;

t1~t2: When  $V_{BS}$  rises above  $UV_{BSR}$ , the circuit does not respond to the input signal immediately, but waits for the next on signal of the input signal;

t2~t3: The circuit works normally;

t3~t4: When  $V_{BS}$  drops to  $UV_{BSD}$  threshold, the circuit enters the undervoltage lockout protection state, and the output is immediately turned off;

t4: When  $V_{BS}$  recovers to above  $UV_{BSR}$ , UVBS enters the reset state, and the circuit works normally from the next on signal of the input signal;

t5: The circuit works normally, the IGBT is on, and the current is provided to the load;

0 ~ t5: Since  $V_{FO}$  only outputs the fault signal of the control power supply undervoltage lockout,  $V_{FO}$  always stays at a high level without fault signal output no matter whether the floating power supply is in the protection state or not.

(See chapter 3, electrical characteristics section for parameter values of <UV>14</UV> and <UV>15</UV>)

#### 4. 4. Description of Overcurrent Protection Function

The LVIC integrates short-circuit protection functionality. It samples the IGBT's operating current through an external current sampling resistor  $R_{CS}$ , with the sampled voltage filtered via RC and connected to port  $C_{SC}$ . When a short circuit occurs, if the voltage at port  $C_{SC}$  exceeds its protection threshold (0.48V), the LVIC activates short-circuit protection. Port  $V_{FO}$  outputs a low-level signal to indicate the fault condition. The gate signal of the low-side IGBT enters soft shutdown mode to prevent circuit failure caused by excessive  $di/dt$ . After complete IGBT shutdown, the short-circuit protection remains active for a duration (no less than 60 $\mu$ s), during which the LVIC remains inactive.

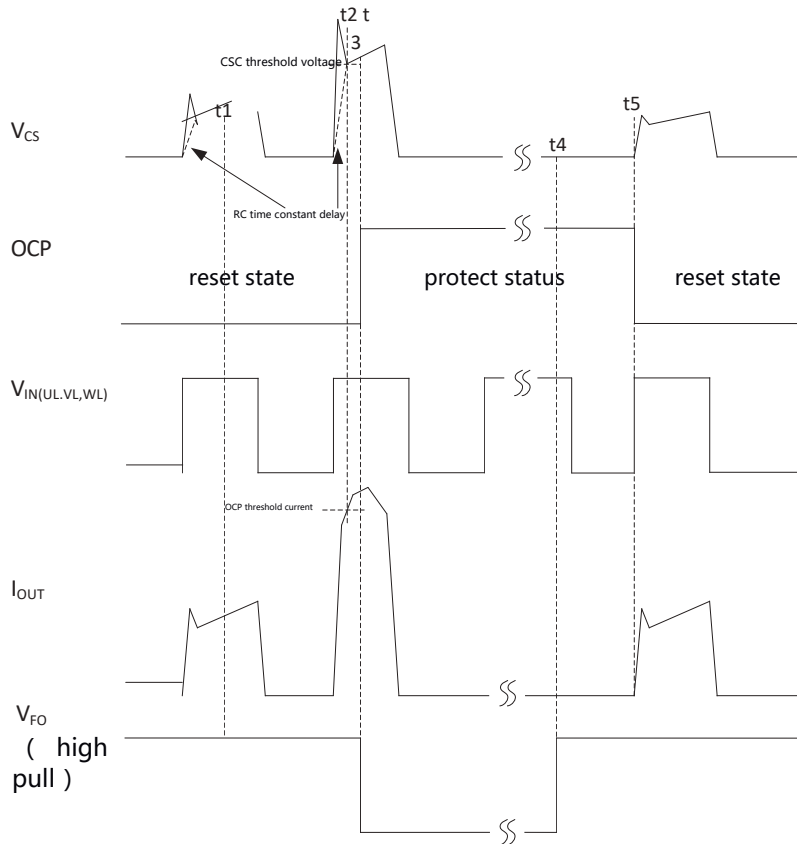


Figure 9 Current Protection Function Timing Diagram

0~t1: The circuit works normally, the IGBT is on, the current is provided to the load,  $V_{FO}$  is always in a high level, and no fault signal is output;

t1~t2: When the OCP voltage monitored on  $V_{CS}$  exceeds the threshold voltage, the overcurrent protection is triggered, but at this time,  $V_{FO}$  is still in a high level and no fault signal is output;

t2~t3: The fault protection module detects the overcurrent fault,  $V_{FO}$  is set to low level, all the gate of the low-side IGBT is hard interrupted, the output is immediately shut down, and the output fault signal is sent;

t3~t4: The fault protection state is maintained for at least 60 $\mu$ s. When the OCP enters the reset state, the circuit starts to work normally from the next on signal of the input signal;

t5: The circuit works normally, the IGBT is on, and the current is provided to the load;

(The  $V_{CS}$  OCP threshold voltage parameter value is shown in Chapter 3,

electrical characteristics)

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5. Typical Application Lines and Application Instructions

5. 1. Application Lines

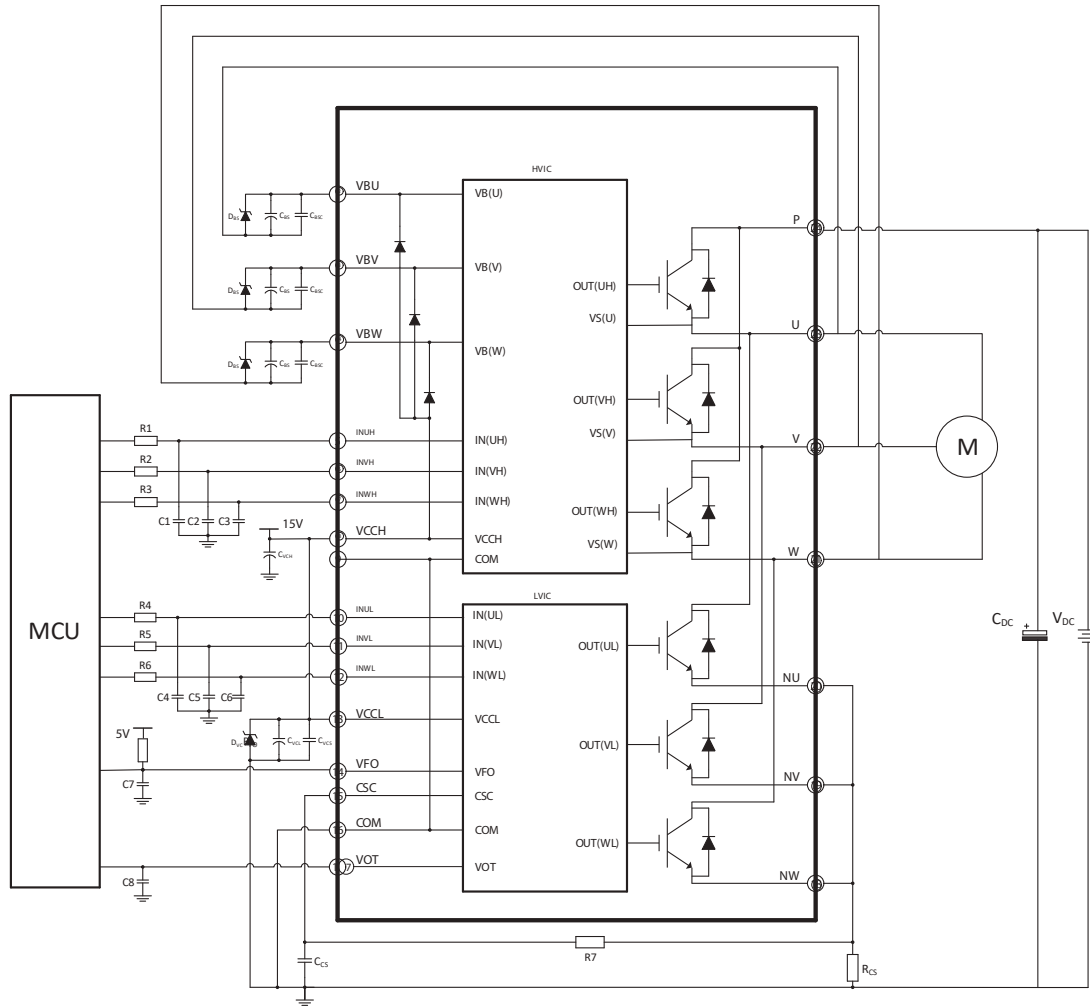


Figure 10 SPB15CDG Typical Application Circuit

5. 2. Application Note

Note 1: the wiring of each input pin should be as short as possible, otherwise it may cause false action;

Note 2: The input signal is high level effective. There is a 5KΩ pull-down resistor connected to ground at the input of each channel in HVIC. In addition, an RC filter circuit can be added at the input end to prevent surge noise caused by incorrect input;

Note 3: In order to prevent surge damage, it is recommended to add a high frequency non-inductive smoothing capacitor (0.1μF~0.22μF) between PN, and the connection of the capacitor should be as short as possible;

Note 4: The connection between the current detection resistor and the IPM should be as short as possible, otherwise the large surge voltage generated by the connecting inductor may cause damage;

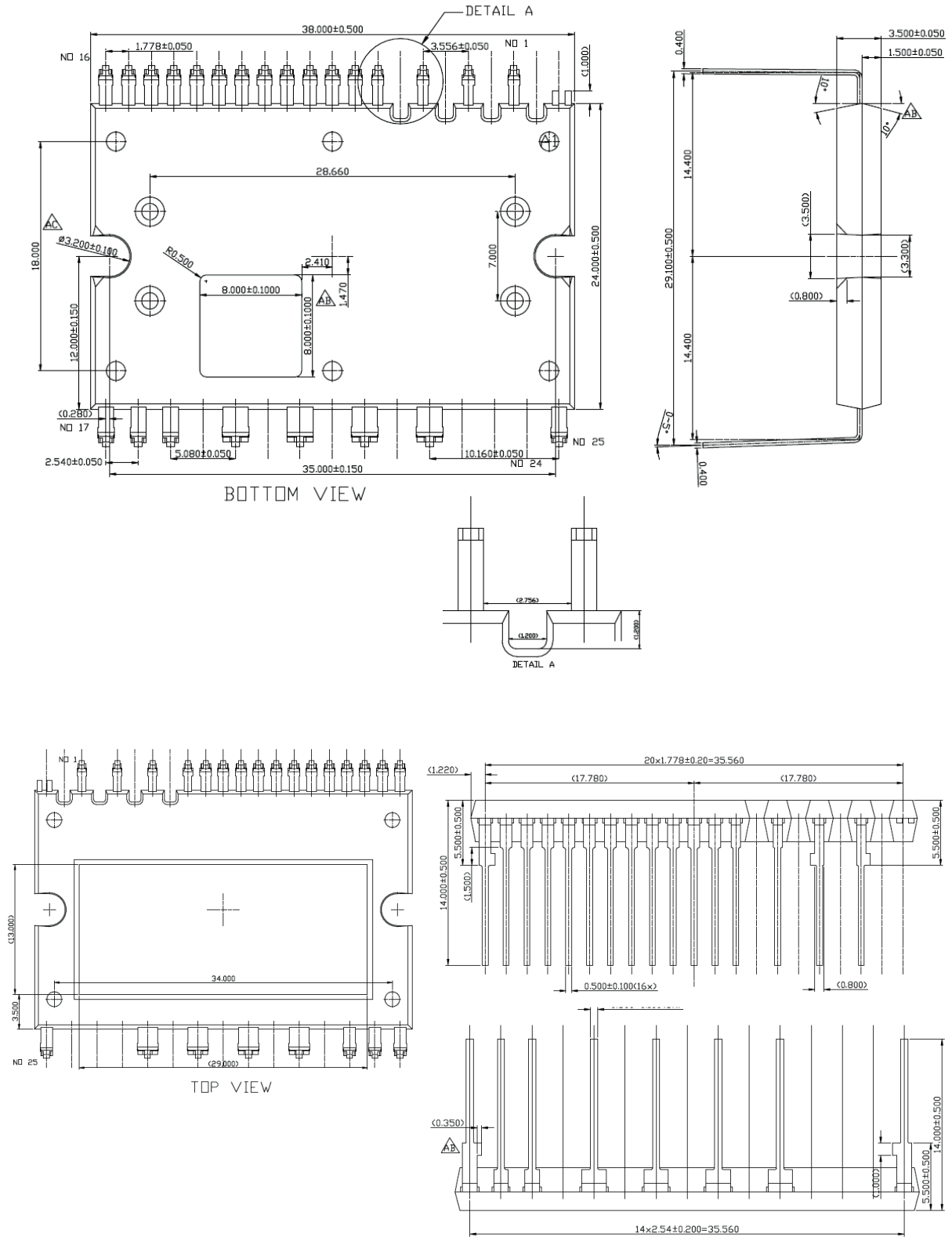
Note 5: The filter capacitor at the 15V power input is recommended to be at least 7 times the self-elevation capacitor CBS;

Note 6: Each external capacitor should be placed as close as possible to the IPM pin;

Note 7: VFO is an open drain output pin, which should be pulled up to 5V power supply through RFO resistor, so that  $I_{FO}$  is 1mA. Do not use it when not in use;

Note 8: In the short circuit protection circuit, please select RCS and CCS with a time constant in the range of 1.5μs to 2 μs, and the wiring around RCS and CCS should be as short as possible, and the wiring of RCS should be close to the shunt resistor.

## 6、 Encapsulation Size and Shape Diagram (unit: Mm)



技术要求  
1、未注公差±0.1

## Description of Toxic and Harmful Substances or Elements in the Product

Name and content of toxic or harmful substances or elements in the product

Part name	Toxic and harmful substances or elements					
	lead (Pb)	mercury (Hg)	cadmium (Cd)	hexavalent chromium (Cr <sup>+6</sup> )	Polybrominated biphenyl (PBB)	Polychlorinated biphenyls (PBDE)
lead frame						
Resin molding						
slug						
intraconnection track						
Film packaging						
explain	: indicates that the content of the toxic and harmful substance is below the limit requirements of GBT26572-2011 standard. ×: indicates that the content of the toxic and harmful substance exceeds the limit requirements of GBT26572-2011 standard.					